# Proton-Assisted Redox-Based Three-Terminal Memristor for Synaptic Device Applications 

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#### Abstract

Emerging technologies, i.e., spintronics, 2D materials, and memristive devices, have been widely investigated as the building block of neuromorphic computing systems. Three-terminal memristor (3TM) is specifically designed to mitigate the challenges encountered by its two-terminal counterpart as it can concurrently execute signal transmission and memory operations. In this work, we present a complementary metal-oxide-semiconductor-compatible 3TM with highly linear weight update characteristics and a dynamic range of $\sim 15$. The switching mechanism is governed by the migration of oxygen ions and protons in and out of the channel under an external gate electric field. The involvement of the protonic defects in the electrochemical reactions is proposed based on the bipolar pulse trains required to initiate the oxidation process and the device electrical characteristics under different humidity levels. For the synaptic operation, an excellent endurance performance with over 256 k synaptic weight updates was  demonstrated while maintaining a stable dynamic range. Additionally, the synaptic performance of the 3 TM is simulated and implemented into a four-layer neural network ( NN ) model, achieving an accuracy of $\sim 92 \%$ in MNIST handwritten digit recognition. With such desirable conductance modulation characteristics, our proposed 3Tmemristor is a promising synaptic device candidate to realize the hardware implementation of the artificial NN.


KEYWORDS: neuromorphic computing, three-terminal, memristor, synaptic device, humidity

## - INTRODUCTION

The complementary metal-oxide-semiconductor (CMOS) technology has been progressively advancing to fulfill the ever-growing demands in computing. The current computational paradigm is based on the conventional von Neumann architecture, dependent on progress in CMOS transistor performance improvement and scaling, in alignment with Moore's law and Dennard's scaling observations. ${ }^{1}$ The von Neumann architecture is designed with separate regions allocated for processing and memory interconnected by data buses. While technological demands have been fulfilled by such an architecture, the growing performance discrepancy between processing and memory units sees an impending limit imposed by performance and scalability. ${ }^{2}$ Neuromorphic computing architecture, capable of parallel data processing, can overcome the von Neumann bottleneck. ${ }^{3,4}$ Furthermore, the biological brain inspires in-memory computing for greater time and energy efficiency. ${ }^{5}$ Extensive studies have been made on the hardware implementation of artificial neural networks (NNs), comprising the fundamental building blocks: artificial neurons and synapses. ${ }^{6,7}$
In the biological brain, the connection strength between neurons is regulated by a synaptic weight. In an artificial synaptic device, analog-resistive switching behavior is used to
mimic the modulation of the synaptic plasticity. ${ }^{8,9}$ Various memristive devices enable analog switching, such as ferroelectric random-access memory, ${ }^{10,11}$ phase-change randomaccess memory, ${ }^{12}$ and resistive random-access memory (RRAM). ${ }^{13}$ These two-terminal (2T) memory technologies have been thoroughly investigated as artificial synapses in prototype NNs. ${ }^{14-17}$ However, 2T synaptic devices face a challenge in simultaneously performing signal transmission and learning function due to the shared writing/reading path. ${ }^{18}$ The shared writing and reading paths would induce extensive noise and disturbance, resulting in reliability issues. ${ }^{19}$ Additionally, 2T memristors normally exhibit a nonlinear conductance change under identical programming pulses, which is not ideal especially for in situ learning applications. A linear conductance modulation (i.e., linear weight updating behavior) enables the direct mapping of the weights in the algorithms to the device conductance states, which leads to

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Figure 1. (a) Schematic and two operation modes of the proposed memristor. (b) Conductance changes during the forming process under bipolar $( \pm 4 \mathrm{~V}, 0.5 \mathrm{~s})$ and negative $(-4 \mathrm{~V}, 0.5 \mathrm{~s})$ gate pulses. Only under the bipolar pulses, the W channel would be electrochemically oxidized. (c) Forming process under different RH values, a higher RH leads to a faster forming process. (d) Forming process under fixed negative pulses ( -4 V , $0.5 \mathrm{~s})$ and varied positive pulse amplitudes $(3 / 3.5 / 4 / 4.5 \mathrm{~V}, 0.5 \mathrm{~s})$. Positive pulses with higher amplitude enable the protons to reach the surface faster, causing faster oxidation.
higher training accuracy. ${ }^{20}$ To overcome these challenges, an additional gate terminal is introduced to decouple the writing/ reading paths and enable more predictable analog conductance update behavior. ${ }^{5,21}$
A 3 T memristor (3TM) can achieve analog conductance modulation when active ions [e.g., $\mathrm{H}^{+}$(proton), $\mathrm{Li}^{+}$, or $\mathrm{O}^{2-}$ ] are doped/de-doped from the electrolyte into the channel. ${ }^{22-24}$ The ion concentration in the channel can be tuned by applying voltage pulses on the gate, while the change in the channel conductance is read out through the drain and source terminal. ${ }^{25}$ Due to the high reactivity of light cations and the challenges they pose on the large-scale integration, investigations on more CMOS compatible oxygen ion based-3TM (O-3TM) have been an emerging topic of interest. ${ }^{9,24,26}$ Recently, O-3TM consisting of $\mathrm{TiO}_{2-x}$ and $\mathrm{WO}_{3}$ switching layers with YSZ electrolytes have demonstrated superior stability with highly linear conductance modulation. ${ }^{5,27}$ In general, the fabrication process of all reported O-3TM devices involves water or other solvents (e.g., lithography, lift-off, etc.) containing moisture/protons. ${ }^{5,27,28}$ Furthermore, the moisture in the test environment can penetrate the device capping layers or side walls, acting as a proton source. Thus, the inherent presence of protons in the device structure becomes inevitable owing to the background moisture during the fabrication process and the device operation even if the device is encapsulated. ${ }^{29}$ However, such impact of humidity on the redox switching of O-3TMs is rarely investigated.
Protons from the moisture in air act as n-type dopants in some oxides channel (e.g., $\mathrm{VO}_{2}$ and $\mathrm{WO}_{x}$ ), more protons in the channel lead to a higher channel conductivity. ${ }^{23,30}$ Meanwhile, the oxides channel deficient/rich in $\mathrm{O}^{2-}$ can also be co-related to high/low conductivity. ${ }^{27}$ In this work, we developed a CMOS-compatible O-3TM consisting of a Pt
layer (gate electrode), a $\mathrm{HfO}_{2}$ layer (electrolyte), and a W layer (channel). To avoid the impact of gate leakage, the high- $\kappa$ dielectric $\mathrm{HfO}_{2}$ with an excellent time-dependent dielectric breakdown performance and good oxygen mobility is considered. ${ }^{24,31-33}$ These materials allowed the implementation of standard established CMOS-fabrication techniques, paving the way for future integration with CMOS circuitry.

The proposed working mechanism of our O-3TM is driven by the migration of protons and $\mathrm{O}^{2-}$ from the channel to the electrolyte or vice-versa under the external gate electric field. It is based on the bipolar pulse trains required to initiate the switching operation (oxidation of the W channel) and the device electrical response under different levels of humidity. The humidity influences the rate of the electrochemical oxidation process of the channel, which is experimentally observed under electrical tests. Such O-3TM demonstrated highly linear channel conductance change behavior, together with stable endurance performance, and high accuracy on pattern recognition simulation based on MNIST data set. All these results suggest that this device has the potential to function as a reliable synaptic element in neuromorphic systems.

## RESULTS AND DISCUSSION

The device comprises an $\mathrm{HfO}_{2}$ insulating layer sandwiched between W and Pt as the channel and the inert electrode, respectively. Depending on the choice of reading and writing paths, the device can be configured into 2 T and 3 T operating modes, as shown in Figure 1a. In the 2T configuration, the shared writing/reading path is indicated by the red arrow. Typical IV characteristics of the $2 \mathrm{~T} \mathrm{Pt} / \mathrm{HfO}_{2} / \mathrm{W}$ device under direct current (DC) voltage sweep are presented in Figure S1. The forming voltage required to initiate the switching is


Figure 2. (a) Gradual activation of the proton transportation in the $\mathrm{HfO}_{2}$ layer under bipolar pulse scheme. Yellow and red arrows indicate the migration direction of protons and $\mathrm{O}^{2-}$, respectively. With positive pulses, the protons move toward the W while $\mathrm{O}^{2-}$ move oppositely. Under negative pulses, the migration direction of proton and $\mathrm{O}^{2-}$ is reversed. Once the protons reach the W surface, the electrochemical oxidation happens with the conductance change indicated in (b) with respective pulse polarities. (c) Structures of the single- and dual-gate devices and the schematic of localized oxidation of the W . This initial oxidation does not occur across the entire $\mathrm{HfO}_{2} / \mathrm{W}$ surface. The red area represents the oxidized W regions.
expected to be high $(\sim 7.0 \mathrm{~V})$ due to the relatively thick $\mathrm{HfO}_{2}$ layer. This forming process induces localized oxygen vacancy defect formation within the RRAM structure. In the process, the compliance current of $100 \mu \mathrm{~A}$ was implemented to prevent the permanent breakdown of the device upon forming from excessive defect generation. The subsequent SET and RESET operations were observed at $\sim 1.3$ and -2.7 V , respectively. These IV characteristics are typically observed in filamentarytype of RRAM devices. ${ }^{34,35}$

Under the 3T operating mode, the writing (blue arrow) and reading (black arrow) paths are decoupled. To initiate the switching operation of the O-3TM, a forming process is required to induce W channel oxidation under an external electric field. Several studies have reported that such a forming step can be achieved by applying a train of negative pulses to the gate, inducing $\mathrm{O}^{2-}$ ion migration toward the channel, resulting in channel oxidation at the interface. This channel oxidation is indicated by the decreasing channel conductance. ${ }^{21,36,37}$ For the developed O-3TM device in this work, the train of negative pulses $(-4 \mathrm{~V}, 0.5 \mathrm{~s})$ failed to oxidize the channel, indicated by the absence of change in channel
conductance (Figure 1b, red). It was found that a bipolar pulse scheme ( $\pm 4 \mathrm{~V}, 0.5 \mathrm{~s}$ ) comprising alternating pulse polarities on the gate was required to induce the electrochemical oxidation process (Figure 1b, blue). Pulse trains applied in this work have the same $1 \mu \mathrm{~s}$ rise/fall time, the pulse interval is varied according to the pulse duration, and the off time between pulses is fixed as 400 ns . The pulse information is given in the form of (pulse amplitude and pulse duration). With consecutive positive pulses, the drop in conductance due to negative pulses becomes larger. This forming process was also found to be dependent on the ambient humidity, as presented in Figure 1c. With the bipolar pulse scheme, the forming process is enhanced as the humidity increases. When pristine devices were placed under relative humidity (RH) of $<20 \%$, the oxidation of W took approximately six times longer than that of a higher humidity condition ( $\mathrm{RH} \sim 70 \%$ ).

It has been reported that the presence of moisture absorbed by the porous oxide layers is essential for the redox reaction in RRAM devices. ${ }^{29,38}$ For our device, the water molecules would first be chemically absorbed on the surface and the grain
(a)

(b)



Low conductance state

Figure 3. (a) $I-V$ characteristics of the 3 T mode. These three hysteresis loops represent the change of $I_{\mathrm{DS}}$ with the applied $V_{\mathrm{GS}}$. Gate current $I_{\mathrm{GS}}$ in the nA range shows a negligible gate leakage current. The arrows indicate the loop direction. (b) Schematic of the channel under high and low conductance states with different amounts of protons and $\mathrm{O}^{2-}$ inside.
boundaries of $\mathrm{HfO}_{2}$, interacting with the oxygen vacancies through the following process

$$
\mathrm{H}_{2} \mathrm{O}+\mathrm{O}_{\mathrm{o}}^{x}+\mathrm{V}_{\mathrm{o}}^{\bullet \bullet} \leftrightharpoons 2 \mathrm{OH}_{\mathrm{o}}^{\bullet}
$$

in which the $\mathrm{OH}_{\mathrm{o}}^{*}$ represents a proton being bonded to an oxygen ion of the metal oxide. ${ }^{39-41}$ The protons from the subsequent physically absorbed water can hop through the $\mathrm{OH}_{\mathrm{o}}^{\bullet}$ network, resulting in protonic conduction governed by the Grotthuss mechanism within the $\mathrm{HfO}_{2}$ layer. ${ }^{39,42}$ As shown in Figure 2a, when the positive pulses arrive, the protons are pumped toward the $\mathrm{HfO}_{2} / \mathrm{W}$ interface while the $\mathrm{O}^{2-}$ in the $\mathrm{HfO}_{2}$ would move in the opposite direction. The hydroxyl groups inside $\mathrm{HfO}_{2}$ allow the protons to travel and ionize the W surface ( $\mathrm{A}, \mathrm{C}$ ). When negative pulses are applied, the $\mathrm{O}^{2-}$ driven to the W surface starts to oxidize the ionized $\mathrm{W}(\mathrm{B}, \mathrm{D})$ with the assistance of the electron acceptor, i.e., proton. Figure 2 b indicates the conductance change under positive and negative pulses. With higher humidity, more hydroxyl groups would be formed, and more protons would reach the W surface, resulting in a faster oxidation process. ${ }^{43}$ Supplementary measurements were conducted with fixed negative pulses ( -4 $\mathrm{V}, 0.5 \mathrm{~s}$ ) but different positive pulses ( $3 / 3.5 / 4 / 4.5 \mathrm{~V}, 0.5 \mathrm{~s}$ ), in which a higher positive pulse enables the protons to reach the W surface faster, leading to a faster oxidation process, as shown in Figure 1d. To confirm that the source of $\mathrm{O}^{2-}$ is the $\mathrm{HfO}_{2}$, XPS depth profiling was performed on a pristine device and the devices under 2 T and 3 T modes after forming operation. The XPS spectra of Hf, O, and W are shown in Figures S2-S5. The spectra can be divided into bulk $\mathrm{HfO}_{2}(<60 \mathrm{~s}), \mathrm{HfO}_{2}-\mathrm{W}$ interface ( $60-70$ s), and bulk W region ( $>70 \mathrm{~s}$ ). The Hf spectra in Figure S2 indicate the presence of different hafnium suboxide states in each device. In the pristine device, $\mathrm{HfO}_{2}$ and $\mathrm{HfO}_{2-x}$ oxide can be found in the bulk region, while the HfO suboxide is found at the $\mathrm{HfO}_{2}-\mathrm{W}$ interface. Meanwhile, the 2 T device shows an additional metallic Hf in the bulk and interface region, suggesting the presence of localized conductive paths within the dielectric. In the 3 T device, suboxides ( $\mathrm{HfO}, \mathrm{HfO}_{2-x}$, and $\mathrm{HfO}_{2}$ ) are present in the bulk and interface, without the metallic Hf. Unlike the soft brokendown of the $\mathrm{HfO}_{2}$ in the 2 T device, $\mathrm{HfO}_{2}$ in the 3 T device remains insulating. This suggests that $\mathrm{O}^{2-}$ are provided by $\mathrm{HfO}_{2}$, and the W channel is effectively oxidized, which can be further confirmed by the W spectra. Additionally, the existence of the metal-hydroxyl interaction and $\mathrm{H}_{2} \mathrm{O}$ in the O 1s spectra of the 3 T device indicates the humidity-promoted oxygen
migration. Unless specifically stated, all the following tests were performed under ambient humidity of RH $\sim 50 \%$.

Another experiment was designed to verify whether this oxidation occurs at the entire channel-oxide interface or in localized regions. Single-gate and dual-gate devices were fabricated with identical device areas (the gate width for the single-gate device is twice that of the dual-gate one), as shown in Figure 2c. The two devices in the dual-gated configuration were formed independently under the same condition as the single-gated devices. After the forming process, it was found that the dual-gate device always has a more insulating channel, indicating non-area-dependent oxidation. The channel resistances for these two devices are present in Table S1. Thus, a localized oxidation model is proposed, as depicted in Figure 2c, in which the red area represents the oxidized W . This localized oxidation promises the scalability potential of the device, which is suitable for high-density synaptic array implementation. To prove its scalability, devices with various channel lengths and widths were fabricated, seeing Figure S6a. The channel length (the area under the $\mathrm{HfO}_{2}$ layer) was varied from 40 to $5 \mu \mathrm{~m}$, and the channel width was varied from 5 to $0.5 \mu \mathrm{~m}$. The forming process of each device was successfully performed, in which the forming time for all devices is around 100 s (Figure $\mathrm{S} 6 \mathrm{~b}-\mathrm{e}$ ). There is no obvious trend for the forming time with the change of the channel dimensions. After the forming process, all devices successfully performed repeatable $P / D$ behaviors, indicating the scalability of the device.

The W channel of each O-3TM was oxidized from $\sim 40$ to $\sim 20 \mathrm{nS}$. DC sweeps on the channel before and after the oxidation was performed, and a transition of the $I-V$ characteristics from linear to nonlinear was observed, as shown in Figure S7a. To further verify the current conduction mechanism of the channel, the nonlinear $I-V$ curve was plotted in the log-log scale and linear fitted (Figure S7b). The slope value of the current increases with the voltage, from 0.997 to 2.654 , suggesting a charge-limited current conduction mechanism. ${ }^{44}$ By fixing the drain reading voltage ( $V_{\mathrm{DS}}$ ) as 0.2 V and dual sweeping the gate voltage $\left(V_{\mathrm{GS}}\right)$ between -1 and 3 V , the change of channel current $\left(I_{\mathrm{DS}}\right)$ and gate current $\left(I_{\mathrm{GS}}\right)$ is shown in Figure 3a. Ionic switching behavior with a large counterclockwise hysteresis loop is observed in the O-3TM. Compared with the channel current (in the $\mu \mathrm{A}$ range), the gate leakage current (in the nA range) is negligible, indicating notable insulating properties of the $\mathrm{HfO}_{2}$ electrolyte. The fluctuation of $\mathrm{I}_{\mathrm{GS}}$ is due to measurement limitations. Similar to the initial oxidation process, the electric field $\left(E_{\text {ext }}\right)$ generated


Figure 4. Transfer and output characteristics of the O-3TM device. (a) Transfer characteristics under different scan rates of the $V_{G S}$ (b) Output characteristics under different amplitude of $V_{\mathrm{GS}}$. (c) Mobile ion migration in the linear region. (d) Mobile ion migration in the saturation region.
by the gate pulse would drive the negatively charged $\mathrm{O}^{2-}$ and positively charged proton in the opposite direction, causing the channel conductance to change. Under positive pulses, the $\mathrm{WO}_{x}$ channel is rich in proton but deficient in $\mathrm{O}^{2-}$, showing a high conductance state. In contrast, more $\mathrm{O}^{2-}$ and less proton in the channel with negative pulses applied would result in a low conductance state. Figure 3b shows the schematic of the $\mathrm{O}-3 \mathrm{TM}$ when the channel is in a high conductance state and a low conductance state. DC sweep of the channel under low and high conductance states is shown in Figure S7c, and the more linear $I-V$ curve of the high conductance state indicates that part of the $\mathrm{WO}_{x}$ channel was reduced back to W by the positive pulses.
The gate-controlled configuration of the $\mathrm{O}-3 \mathrm{TM}$ is very similar to a transistor, with the semiconductor channel replaced by the $\mathrm{WO}_{x}$ layer. With the transfer and output characteristics given in Figure 4, an n-type enhancement-mode transistor-like performance can be observed in the O-3TM. ${ }^{45}$ Figure 4a shows the transfer curves under different scan rates of $V_{\mathrm{GS}}$. A slower scan rate correlates to a bigger hysteresis loop and a larger high/low current ratio when $V_{\mathrm{GS}}=0 \mathrm{~V}$. With a faster scan rate, a smaller amount of mobile ions would be migrated, resulting in a smaller current change. In comparison, a slower scan rate allows more mobile ions to be activated and migrated, contributing to a more pronounced current change. The output characteristics were obtained by sweeping the $V_{\mathrm{DS}}$ from 0 to 3 V at different $V_{\mathrm{GS}}$, as shown in Figure 4b. Like a MOSFET, the device also exhibits linear and saturation regions. For a MOSFET, the saturation region appears when the inversion layer starts to pinch off due to the high drain voltage. Nevertheless, the saturation region for the O-3TM can be assigned to the competition between different electric fields. With positive $V_{\mathrm{GS}}$ applied, the $E_{\text {ext }}$ pumps the mobile ions to
migrate accordingly until the channel is depleted of $\mathrm{O}^{2-}$ and rich in the proton. At the same time, the concentration gradient of $\mathrm{O}^{2-}$ and proton would generate an increasing internal electric field $\left(E_{\text {int }}\right)$ that has the opposite direction to the $E_{\text {ext }}$. The applied $V_{\mathrm{DS}}$ also introduces an electric field $\left(E_{\mathrm{D}}\right)$ near the drain terminal. In the linear region, the $E_{\text {ext }}$ overwhelms $E_{\text {int }}$ and $E_{\mathrm{D}}$ and leads to a monotonic rise in conductance (Figure 4c). With the increase of $E_{D}$ and the strong $E_{\text {int }}$ introduced by high $V_{\mathrm{GS}}$, the $E_{\text {ext }}$ cannot keep modulating the ion concentration of the channel, leading to the saturation region (Figure 4d).

With the $\mathrm{O}^{2-}$ /proton migration in and out of the $\mathrm{WO}_{x}$ channel, this O-3TM can be operated as an artificial synapse. ${ }^{9,27}$ The synaptic weight is represented by the channel conductance, which can be tuned by the applied gate pulses, thus emulating the weight updating in biological synapses. The increase or decrease of synaptic weight is called potentiation $(P)$ and depression (D). For artificial synapses, a linear, symmetrical, and multi-bit $P / D$ behavior with a large dynamic range is critical for the training accuracy of a NN. ${ }^{12,27}$ A model is used to evaluate the nonlinearity of the $P / D$ process (see eqs S1-S3). ${ }^{45,46}$ For the ideal $P / D$ curve, the nonlinearity factor " $A$ " value is 0 .

For the developed O-3TM device, the synaptic operations, i.e., potentiation and depression, are performed by applying positive and negative voltage pulses on the gate, respectively. Figure S8a shows the $P / D$ curves with 20 potentiation pulses $(2.8 \mathrm{~V}, 50 \mathrm{~ms})$ followed by 20 depression pulses $(-0.8 \mathrm{~V}, 50$ ms ) under a fixed read $V_{\mathrm{DS}}$ of 0.2 V . The potentiation curve achieved a nonlinearity factor of 2.40 , while significantly less linear depression behavior with a nonlinearity factor of 7.16 was obtained. The linearity of the depression curve is improved as the negative voltage amplitude of the depression is reduced,


Figure 5. Measured channel conductance modulation of O-3TM by applying gate pulses. (a) $P / D$ curves under different RH. Under a higher humidity condition, the conductance window becomes wider, in addition to a higher minimum conductance state. (b) P/D curves of the O-3TM were obtained by applying $256 \mathrm{P} / \mathrm{D}$ programming pulses with an amplitude of $3 / 0 \mathrm{~V}$ and a duration of 5 ms . The $V_{\mathrm{DS}}$ is fixed as 0.2 V to read out the channel conductance. (c) Nonlinearity analysis of the $P / D$ curves with $A_{\mathrm{P}}$ of 0.98 and $A_{\mathrm{D}}$ of 0.44 indicates quasi-linear and symmetry weight update behavior. (d) Conductance changes after each writing pulse ( $3 \mathrm{~V}, 50 \mathrm{~ms}$, space 1 s ). Reading time of 1 s with no voltage applied to monitor the conductance states between two pulses. (e) Potentiation pulse-amplitude-dependent $P / D$ curves of the O-3TM. Pulse duration is fixed as 5 ms. $\Delta G$ is exponentially dependent on the pulse width (inset). (f) Potentiation pulses-width-dependent $P / D$ curves of the $\mathrm{O}-3 \mathrm{TM} . P / D$ pulse amplitude is fixed as $3 / 0 \mathrm{~V} . \Delta G$ increases linearly with the pulse width (inset).
as depicted in Figure S8a-d. The smallest nonlinearity factor was achieved under 0 V in which no external gate pulses were applied, and the conductance state was read every 50 ms to match the potentiation duration. This voltage-dependent depression performance indicates that the ideal depression behavior is governed by the self-diffusion of the $\mathrm{O}^{2-}$ /proton mechanism rather than applying an external electric field to drive them.

First, the $P / D$ behavior under different humidity was investigated by sequentially programming the O-3TM with 256 potentiation pulses ( $2.8 \mathrm{~V}, 5 \mathrm{~ms}$ ) followed by 256 depression pulses ( $0 \mathrm{~V}, 5 \mathrm{~ms}$ ) and fixing the $V_{\mathrm{DS}}$ as 0.2 V . The device presents a wider conductance window and a higher minimum conductance state with the increase of humidity, as shown in Figure 5a. As more protons migrate toward the channel and participate as charge carriers in a more humid condition, a more pronounced conductance change could be observed. Under opposite pulse polarities, the channel conductance increases or decreases due to the movement of protons and $\mathrm{O}^{2-}$. However, to simplify the description of the ion migration, we will only focus on the migration of $\mathrm{O}^{2-}$ in the subsequent discussion.

When writing the device with $2563 / 0 \mathrm{~V}(5 \mathrm{~ms})$ pulses, the device features good linearity of $\mathrm{P} / \mathrm{D}$ with a large conductance dynamic range of $\sim 15$ under ambient humidity (Figure 5b). As Figure 5 c shows, the nonlinearity factor of potentiation $\left(A_{\mathrm{P}}\right)$ and depression $\left(A_{\mathrm{D}}\right)$ is 0.98 and 0.44 , respectively, indicating a quasi-linear conductance update behavior. In contrast, the $P / D$ curves of the device under 2 T operation mode are nonlinear and unpredictable ( $A_{\mathrm{P}}=10.68, A_{\mathrm{D}}=13.43$, Figure S9). Figure 5 d reveals the conductance change $(\Delta G)$ after each writing
pulse ( $3 \mathrm{~V}, 50 \mathrm{~ms}$ ). In between each pulse, the conductance was monitored for 1 s . As shown in Figure 5d, for higher conductance states, more $\mathrm{O}^{2-}$ are extracted out of the channel, which correlates to a higher $E_{\text {int }}$ so the diffusion of $\mathrm{O}^{2-}$ back into the channel will be enhanced after the removal of $E_{\text {ext }}$. A small positive voltage of 0.8 V can be applied to hold each conductance state, and the retention of 1000 s for 10 chosen conductance states is shown in Figure S10.

For the O-3TM, the applied gate pulses with different amplitude or width would extract the $\mathrm{O}^{2-}$ to varying degrees, thus leading to different $P / D$ behaviors. It is necessary to inspect the $P / D$ behaviors under different pulse conditions to further investigate the O-3TM programmability. Potentiation pulse-amplitude-dependent and pulses-width-dependent $P / D$ curves are shown in Figure 5e,f, in which the conductance value of each curve is normalized to show a more intuitive trend. In Figure 5e, $P / D$ curves show an increased dynamic range when the potentiation pulse amplitude increases from 2.6 to 3.4 V (pulse duration was fixed as 5 ms ), in which the inset indicates that $\Delta G$ is exponentially dependent on the pulse amplitude. Figure 5 f shows the $P / D$ curves with the potentiation pulse-width varying from 2 to 50 ms under fixed pulse amplitude as 3 V . The dynamic range increases with the pulse width while the linearity degrades instead. When the pulse width is 20 ms and above, the conductance increases or decreases rapidly at first but becomes saturated at the end of the process. The inset in Figure 5f shows that the $\Delta G$ has an approximately linear relationship with the pulse width. It still offers a distinctive $\Delta G$ of 47 nS when the pulse width is decreased to 2 ms , demonstrating a high-speed switching compared with reported oxygen-ion based synaptic devices. ${ }^{27}$


Figure 6. (a) Fast-switching speed of $100 \mu$ s is demonstrated in the O-3TM. (b) Excellent endurance performance of over 256k pulses, $P / D$ pulses were fixed as $3 / 0 \mathrm{~V}, 5 \mathrm{~ms}$. (c) Cycle-to-cycle and (d) device-to-device variation of the $\mathrm{O}-3 \mathrm{TM}$, in which the red line indicates the average value, and the gray error bars are the respective standard deviation.


Figure 7. (a) Schematic of the four-layer neuron network for the pattern recognition simulation architecture. Each node represents the neuron, and each connection represents the synapse. (b) Pattern recognition accuracies based on the ideal synaptic device and the O-3TM.

The switching speed of synaptic devices is a vital characteristic for online training of NN $s$, which is determined by the duration of the programming pulse. ${ }^{47}$

To further investigate the speed limit of the O-3TM, it is programmed with 20 potentiation pulses and 20 depression pulses with higher amplitude and shorter pulse width. Conductance modulation of the O-3TM is demonstrated with a pulse width of $100 \mu$ s (Figure 6a), indicating that a fastswitching speed is possible in the O-3TM. For online training, the weights of each synapse are updated frequently, which puts forward requirements on the endurance performance of synapses. ${ }^{20}$ Strong endurance characteristics are demonstrated in Figure 6b by repeatedly programming the device with 256 k
pulses. More than 500 stable $P / D$ cycles can be achieved with 256 conductance updates for each $P / D$ process. As the device underwent more $P / D$ cycles, the conductance started to decrease and eventually stuck in the lower conductance state (a significant drop can be observed after $2000 P / D$ cycles). This is potentially due to the expansion of the oxidized W region during device operation, which leads to fully oxidized W channel. Except for the speed and endurance requirements, small device-to-device and cycle-to-cycle variations are also crucial for high training accuracy. ${ }^{48}$ Figure 6 c shows the cycle-to-cycle variation of 100 cycles. A steady operation without significant degradation of the $P / D$ performance is observed with an average variation of $3.6 \%$. Besides, the device-to-device
variation was obtained from five devices with an average variation of $18.9 \%$, as shown in Figure 6d. The variation for the potentiation is a bit higher than that of the depression process, which can be attributed to the different oxidation degrees of the W channel during the initial oxidation process. Such a low variation could empower the O-3TM to implement the NN as the synapse with higher stability. Error bars in Figure 6 represent the standard deviation of the conductance update, and the variation is calculated by dividing the standard variation by the averaged conductance value.
In order to evaluate the performance of the O-3TM as an artificial synapse, a pattern recognition simulation was conducted within a four-layer NN. The NN consists of one input layer ( 784 neurons), two hidden layers ( 250 and 125 neurons), and one output layer ( 10 neurons), in which every two neurons are connected by synapses (Figure 7a). ${ }^{49}$ Patterns for the recognition training are handwritten digits ( 0 to 9 ) from the Modified National Institute of Standards and Technology (MNIST) database with $28 \times 28$ pixels. For each epoch, 50,000 images from the MNIST training set and 10,000 images from the MNIST test set are used for training and testing, respectively. ${ }^{50}$ During the training process, the weight of each synapse is being updated according to the linearity, dynamic range, and precision extracted from Figure 5 b . After 30 training epochs, the high recognition accuracy of $92.36 \%$ can be achieved by the O-3TM (Figure 7b). The same simulation was conducted using an ideal synaptic device with perfect linearity ( $A_{\mathrm{P}}=A_{\mathrm{D}}=0$ ), infinite dynamic range, and infinite conductance states, achieving 97.07\% accuracy. Compared with the ideal case, the $\mathrm{O}-3 \mathrm{TM}$ is an up-andcoming candidate as the artificial synapse for neuromorphic computing.

## - CONCLUSIONS

In summary, a CMOS compatible $\mathrm{O}-3 \mathrm{TM}$ based on $\mathrm{O}^{2-}$ / proton migration has been developed as an artificial synapse. It shows nearly linear potentiation ( $A_{\mathrm{P}}=0.98$ ) and depression $\left(A_{\mathrm{D}}=0.44\right)$ behavior under identical pulses operation, while maintaining a stable dynamic range of $\sim 15$. The initial electrochemical oxidation process is found to be dependent on the humidity level. The higher oxidation rate is observed under higher humidity because of the proton-assisted mechanism. In addition, the electrochemical oxidation of the W during the forming process is not area dependent, thus providing the scalability potential on the device level. A relatively fast switching speed of $100 \mu \mathrm{~s}$ and high endurance of more than 256 k weight updates were also demonstrated in this O-3TM. The high MNIST pattern recognition accuracy of $92.36 \%$ achieved by the O-3TM indicates its capability to act as an artificial synapse in the next generation of the neuromorphic system.

## ■ EXPERIMENTAL METHODS

The O-3TM device was fabricated on a $\mathrm{Si}\langle 100\rangle$ substrate with 300 nm thermal oxide above. Before the deposition of each layer, a UV lithography process was conducted to form the desired pattern. Each lithography process started with the spin coating of AZ 5214 E photoresist, followed by the soft baking for 90 s at $115{ }^{\circ} \mathrm{C}$ and exposure under the SUSS Microtec MJB4 mask aligner for 13 s with the patterned mask. First, the 5 nm thick W channel layer was deposited by DC sputtering under power conditions of 50 W . Second, a 30 nm thick $\mathrm{HfO}_{2}$ layer was deposited by radio frequency (RF) sputtering, followed by the 15 nm Pt top electrode deposition by DC sputtering, both under power conditions of 50 W . Lastly, a 20 nm
thick Pt was deposited as source/drain contact pads with the size of $150 \times 150 \mu \mathrm{~m}^{2}$. All sputtering processes were kept at 2 mTorr under Ar ambient at room temperature using the AJA International ATCOrion 8 -target UHV magnetron sputtering deposition system. By using the Keithley 4200A SCS parameter analyzer, the electrical properties and synaptic behaviors of the device were measured in an ambient environment at room temperature.

## ASSOCIATED CONTENT

## (s) Supporting Information

The Supporting Information is available free of charge at https://pubs.acs.org/doi/10.1021/acsami.3c03974.

Bipolar $I-V$ characteristics of the $2 \mathrm{~T} \mathrm{Pt} / \mathrm{HfO}_{2} / \mathrm{W}$ devices; channel resistance for single- and dual-gate device after the oxidation; XPS spectra of Hf 4 f for pristine, 2 T , and 3 T devices under different etching cycles; XPS spectra of O 1s for pristine, 2 T , and 3 T devices under different etching cycles; XPS spectra of W 4 f for pristine, 2 T , and 3 T devices under different etching cycles; etching time-atomic concentration profiles for $\mathrm{W}, \mathrm{Hf}$, and O elements in pristine, 2 T , and 3 T devices; images of devices with varied channel lengths and widths and respective forming processes; DC sweep of the channel and conduction mechanism fitting; $P / D$ curves under fixed varies depression voltage; $P / D$ curves of the device under 2 T operation mode; 10 conductance states retention under 0.8 V holding voltage; and nonlinearity analysis model (PDF)

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## Notes

The authors declare no competing financial interest.

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